

Fundamentals of Electronic Circuits and Systems I

# Wrap Up

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Physics

Foundations of  
Electronic Circuits & Systems

## L1 : LMD

- The rate of change of magnetic flux linked with any portion of the circuit must be zero for all time.

$$\oint E \cdot dl = -\frac{\partial \Phi_B}{\partial t} = 0$$

- The rate of change of the charge at any node in the circuit must be zero for all time. A node is any point in the circuit at which two or more element terminals are connected using wires.

$$\int_{S_x} J \cdot dS - \int_{S_y} J \cdot dS = \frac{\partial q}{\partial t} = 0$$

- The signal timescales must be much larger than the propagation delay of electromagnetic waves through the circuit.

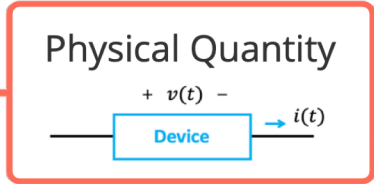
**LMD:**

**The fundamental of circuit and system theory**



**L1 : LMD**

# Foundations of Electronic Circuits & Systems



Electronic Devices

Circuit Analysis Skill

The rate of change of magnetic flux linked with any portion of the circuit must be zero for all time.

$$\oint E \cdot dl = -\frac{\partial \Phi_m}{\partial t} = 0$$

The rate of change of the charge at any node in the circuit must be zero for all time. A node is any point in the circuit at which two or more element terminals are connected using wires.

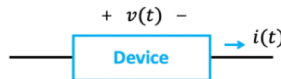
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**LMD:**  
The fundamental of circuit and system theory

# Foundations of Electronic Circuits & Systems

## Physical Quantity



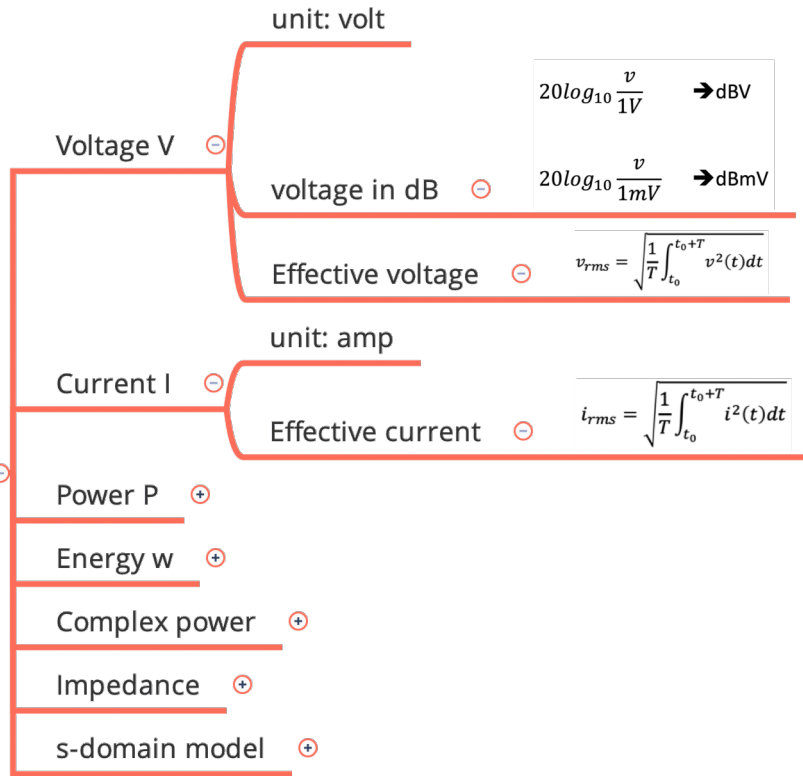
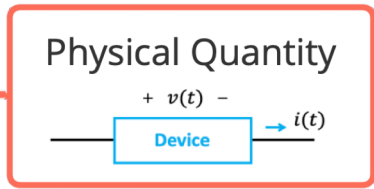
- Voltage  $V$  ⊕
- Current  $I$  ⊕
- Power  $P$  ⊕
- Energy  $w$  ⊕
- Complex power ⊕
- Impedance ⊕
- s-domain model ⊕

## Electronic Devices

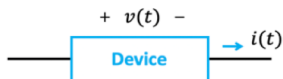
- Linear v.s. Non-linear ⊕
- passive v.s. active ⊕

## Circuit Analysis Skill

- Kirchhoff's Law ⊕
- Linear v.s. Non-linear ⊕
- Time v.s. Frequency ⊕
- Feedback ⊕
- Additional skills ⊕



## Physical Quantity



Voltage V ⊕

Current I ⊕

unit: watt

Instantaneous power ⊖  $p(t) = v(t)i(t)$

Average power ⊖  $\bar{p} = \frac{1}{T} \int_{t_0}^{t_0+T} p(t) dt$

Power P ⊖

Maximum Power Transfer ⊖

Energy w ⊖ unit: Joule ⊖  $w = \int_0^T p(t) dt$

Complex power ⊕

Impedance ⊕

s-domain model ⊕

A circuit diagram showing a battery symbol with a plus sign on top and a minus sign on bottom, labeled  $v_s$ . To its right is a resistor labeled  $R_s$ . These two are enclosed in a dashed box. To the right of this dashed box is another resistor labeled  $R_L$ . An arrow labeled  $i$  indicates current flowing from the positive terminal of the source, through  $R_s$ , and then through  $R_L$ .

**Practical voltage source**

- Voltage of  $R_L$   
 $v_{R_L} = v_s - iR_s$
- Power at the load  $R_L$   
$$P_L = v_{R_L} i = (v_s - iR_s) i = -R_s \left( i^2 - \frac{v_s}{R_s} i \right) = -R_s \left( i - \frac{1}{2} \frac{v_s}{R_s} \right)^2 + \frac{1}{4} \frac{v_s^2}{R_s}$$
  
$$\leq \frac{1}{4} \frac{v_{s,rms}^2}{R_s}$$
 **The maximum power being absorbed by the load**
- When  $R_s = R_L$   $P_L = P_{L,max} = \frac{1}{4} \frac{v_{s,rms}^2}{R_s}$

Definition  $\mathbf{S} = \frac{1}{2} \mathbf{VI}^*$

unit: watt

Average power

$\bar{P} = \Re\{\mathbf{S}\} = V_{rms} I_{rms} \cos(\varphi_V - \varphi_I)$

unit: Volt-Ampere Reactive/VAR

Reactive power

$Q = \Im\{\mathbf{S}\} = j V_{rms} I_{rms} \sin(\varphi_V - \varphi_I)$

power factor

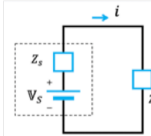
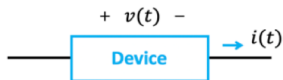
$pf = \frac{P}{V_{rms} I_{rms}} = \cos(\varphi_V - \varphi_I)$

power angle

$\tan(\varphi_V - \varphi_I) = \frac{Q}{P}$

Complex power

Physical Quantity



$$\begin{cases} I_L = \frac{V_s}{\sqrt{(R_s + R_L)^2 + (X_s + X_L)^2}} \\ V_L = \frac{V_s \sqrt{R_L^2 + X_L^2}}{\sqrt{(R_s + R_L)^2 + (X_s + X_L)^2}} \end{cases}$$

The average power at the load  $Z_L$

$$P_L = V_{rms} I_{rms} \cos(\theta_{v_L} - \theta_{i_L}) \quad \text{where } \theta_{v_L} - \theta_{i_L} = \cos^{-1}\left(\frac{R_L}{\sqrt{R_L^2 + X_L^2}}\right)$$

$$= \frac{V_{rms}^2 R_L}{(R_s + R_L)^2 + (X_s + X_L)^2} \leq \frac{V_{rms}^2 R_L}{(R_s + R_L)^2} \leq \frac{1}{4} \frac{V_{rms}^2}{R_s}$$

Maximum Power Transfer

when  $\begin{cases} R_s = R_L \\ X_s = -X_L \end{cases} \rightarrow Z_s = Z_L^* \quad P_{L,max} = \frac{1}{4} \frac{V_{rms}^2}{R_s}$

Impedance is defined as the ratio of the phasor voltage to the phasor current

Impedance

<b>i-v relation</b>	$i = \frac{v}{R}$	$i(t) = C \frac{dv(t)}{dt}$	$i(t) = i(t_0) + \frac{1}{L} \int_{t_0}^t v(x) dx$
<b>v-i relation</b>	$v = iR$	$v(t) = v(t_0) + \frac{1}{C} \int_{t_0}^t i(x) dx$	$v(t) = L \frac{di(t)}{dt}$
<b>Impedance</b>	$R$	$\frac{1}{j\omega C}$	$j\omega L$

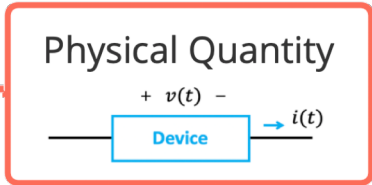
s-domain model

based on Laplace transform  
(WILL LEARN IN SIGNAL & SYSTEM)

<b>s-domain model</b>	$R$	$\frac{1}{sC}$	$sL$
<b>s-domain i-v relation</b>	$I(s) = \frac{V(s)}{R}$	$I(s) = sCV(s)$	$I(s) = \frac{V(s)}{sL}$

Linear v.s. Non-linear

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- Voltage V ⊕
- Current I ⊕
- Power P ⊕
- Energy w ⊕
- Complex power ⊕
- Impedance ⊕
- s-domain model ⊕

Additivity:  
response to SUM OF INPUTS is the SUM OF THE RESPONSES to each input applied separately

Homogeneity:  
if input (excitation) is multiplied by a constant, then output (response) is multiplied by the SAME constant

A device is LINEAR if it satisfies 2 properties:

Linear v.s. Non-linear ⊖

Electronic Devices ⊖

passive v.s. active ⊖

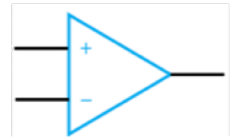
Passive devices ⊖

- Resistor ⊕
- Capacitor ⊕
- Inductor ⊕
- Comparison ⊕
- Mutual inductance ⊕

Active devices ⊖

power source ⊕

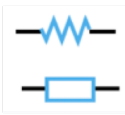
Op-Amp = Operational Amplifier



Kirchhoff's Law ⊕



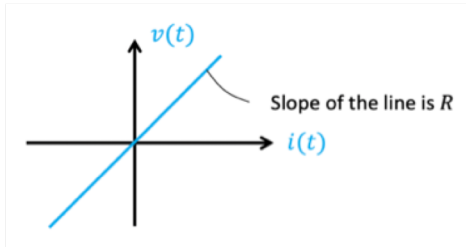
Resistance represents the ability to resist flow of electric current



Ohm's law: the voltage across a resistance is directly proportional to the current flowing through it.

$$v(t) = Ri(t)$$

I-V relationship of ideal resistor



Resistor +

Capacitor +

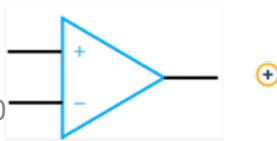
Inductor +

Comparison +

Mutual inductance +

power source +

Op-Amp =  
Operational  
Amplifier



A CAPACITOR is a circuit element that consists of two conducting surface separated by a non-conducting, or dielectric, material.



unit: farad (F)

I-V relationship  $i(t) = C \frac{dv(t)}{dt}$

Voltage on capacitor CANNOT change abruptly

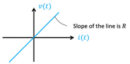
An capacitor is an open circuit at DC

Resistance represents the ability to resist flow of electric current



Ohm's law: the voltage across a resistance is directly proportional to the current flowing through it.  $v(t) = R i(t)$

I-V relationship of ideal resistor



Resistor +

Capacitor +

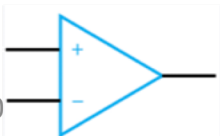
Inductor +

Comparison +

Mutual inductance +

power source +

Op-Amp = Operational Amplifier +



An INDUCTOR is a circuit element that consists of a conducting wire usually in the form of a coil



I-V relationship  $v(t) = L \frac{di(t)}{dt}$

unit: henry (H)

Current through inductor CANNOT change instantaneously

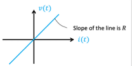
An inductor is a short circuit at DC

Resistance represents the ability to resist flow of electric current



Ohm's law: the voltage across a resistance is directly proportional to the current flowing through it.  $v(t) = R i(t)$

I-V relationship of ideal resistor



A CAPACITOR is a circuit element that consists of two conducting surface separated by a non-conducting, or dielectric, material.  $\frac{+}{-} c$

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Resistor +

Capacitor +

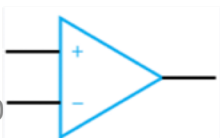
Inductor +

Comparison +

Mutual inductance +

power source +

Op-Amp = Operational Amplifier



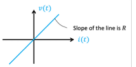
+

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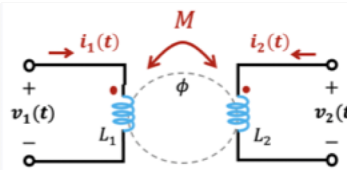
Mutual inductance +

power source +

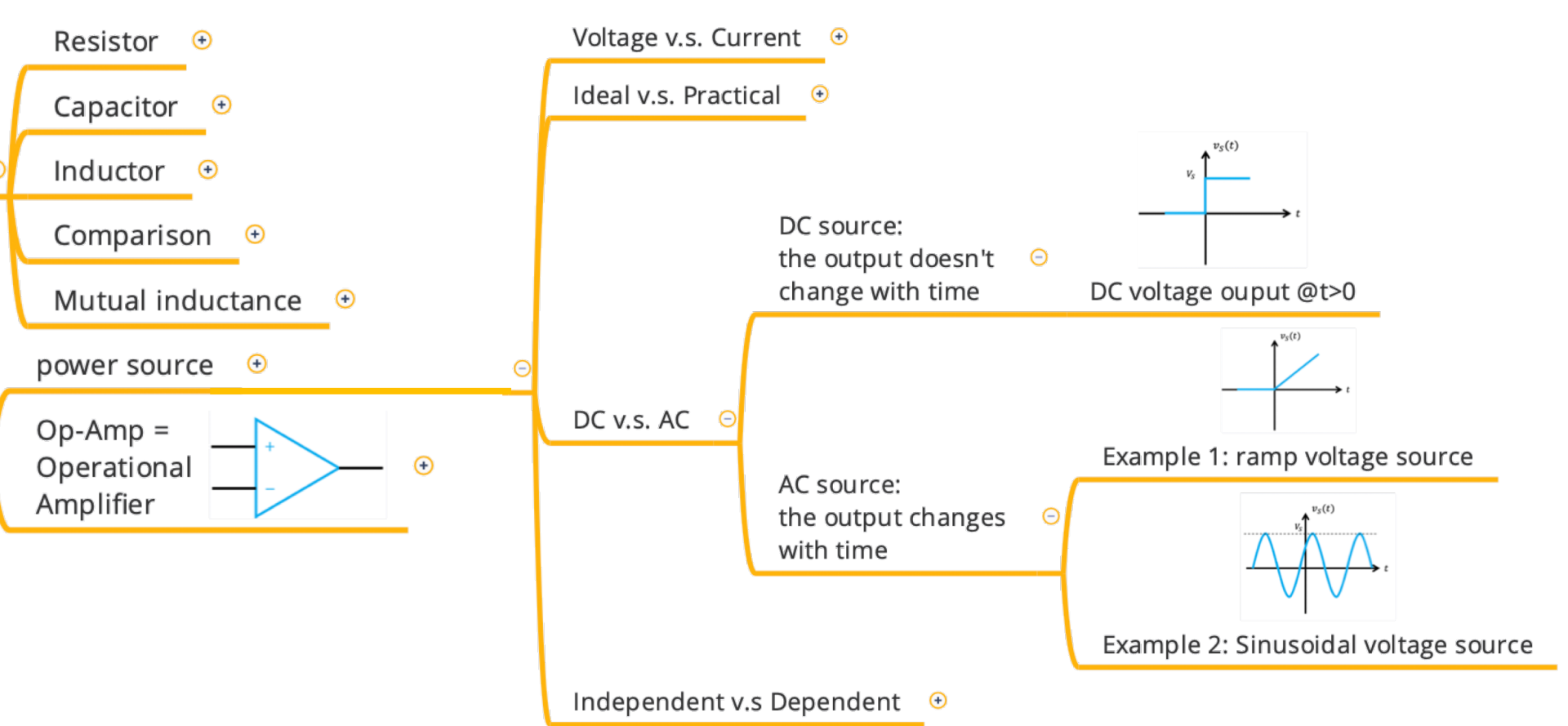


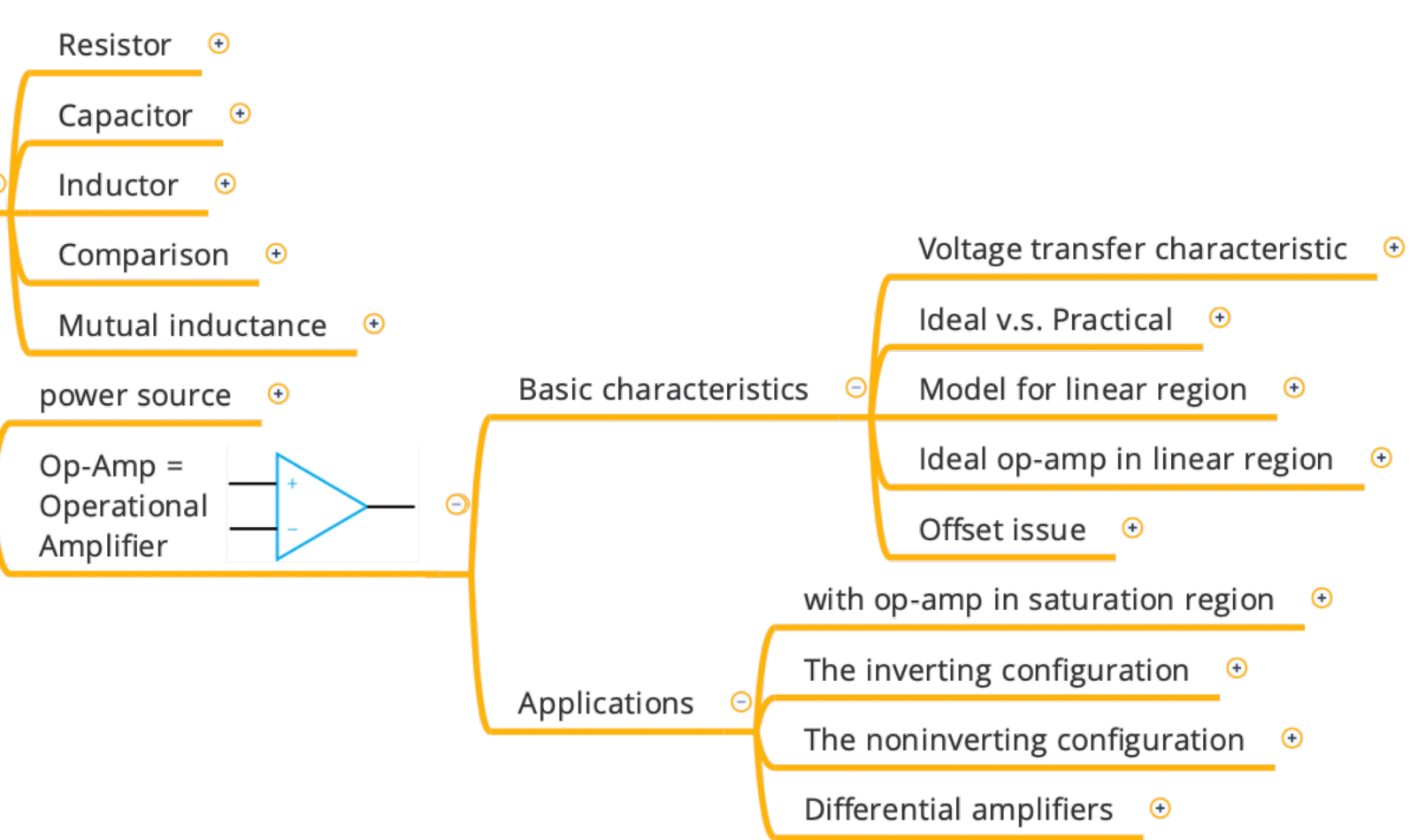
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<b><i>p</i> (power transferred in)</b>	$p = vi$	$p = vi$	$p = vi$
<b><i>w</i> (stored energy)</b>	0	$w = \frac{1}{2} C v^2(t)$	$w = \frac{1}{2} L i^2(t)$
<b>Series combination</b>	$R_{eq} = R_1 + R_2$	$C_{eq} = \frac{C_1 C_2}{C_1 + C_2}$	$L_{eq} = L_1 + L_2$
<b>Parallel combination</b>	$R_{eq} = \frac{R_1 R_2}{R_1 + R_2}$	$C_{eq} = C_1 + C_2$	$L_{eq} = \frac{L_1 L_2}{L_1 + L_2}$
<b>DC behavior</b>	NO	open circuit	short circuit
<b>Instantaneous change of <i>v</i></b>	✓	×	✓
<b>Instantaneous change of <i>i</i></b>	✓	✓	×

I-V relationship



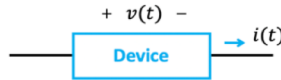
$$\begin{cases} v_1 = L_1 \frac{di_1}{dt} + M \frac{di_2}{dt} \\ v_2 = M \frac{di_1}{dt} + L_2 \frac{di_2}{dt} \end{cases}$$





# Foundations of Electronic Circuits & Systems

## Physical Quantity



- Voltage  $V$  ⊕
- Current  $I$  ⊕
- Power  $P$  ⊕
- Energy  $w$  ⊕
- Complex power ⊕
- Impedance ⊕
- s-domain model ⊕

## Electronic Devices

- Linear v.s. Non-linear ⊕
- passive v.s. active ⊕

## Circuit Analysis Skill

- Kirchhoff's Law ⊕
- Linear v.s. Non-linear ⊕
- Time v.s. Frequency ⊕
- Feedback ⊕
- Additional skills ⊕

## Circuit Analysis Skill

Kirchhoff's Law

Linear v.s. Non-linear

Time v.s. Frequency

Feedback

Additional skills

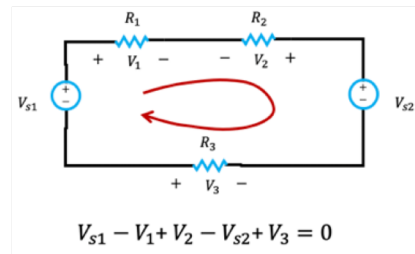
KVL: Sum of voltages  
around a closed  
path is zero

$$\sum_{n=1}^N v_n = 0$$

KCL: Sum of currents  
entering a node is zero

$$\sum_{n=1}^N i_n = 0$$





KVL: Sum of voltages around a closed path is zero

$$\sum_{n=1}^N v_n = 0$$

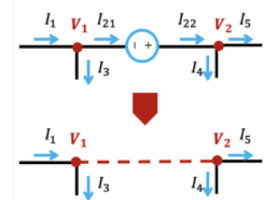
KCL: Sum of currents entering a node is zero

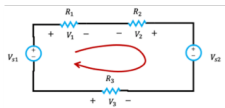
$$\sum_{n=1}^N i_n = 0$$

node voltage

- Step 1a: identify all extraordinary nodes
- Step 1b: Assign ground node
- Step 1c: assign node voltages to the rest
- Step 2a: perform KCL at each node
- Step 2b: apply ohm's law for each current
- Step 3: solve the system of equations

a SUPERNODE is formed when a voltage source connects two extraordinary nodes



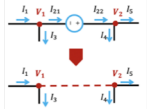


$$V_{s1} - V_1 + V_2 - V_{s2} + V_3 = 0$$

- Step 1a: identify all extraordinary nodes
- Step 1b: Assign ground node
- Step 1c: assign node voltages to the rest
- Step 2a: perform KCL at each node
- Step 2b: apply ohm's law for each current
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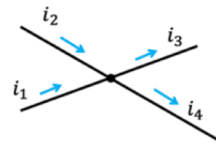
node voltage

a SUPERNODE is formed when a voltage source connects two extraordinary nodes



KVL: Sum of voltages around a closed path is zero

$$\sum_{n=1}^N v_n = 0$$



$$i_1 + i_2 - i_3 - i_4 = 0$$

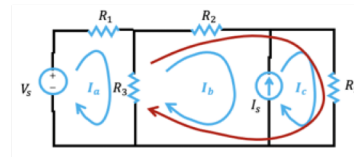
KCL: Sum of currents entering a node is zero

$$\sum_{n=1}^N i_n = 0$$

- Step 1a: identify all mesh
- Step 1b: assign each mesh an unknown current, usually in clockwise direction
- Step 2a: apply KVL to each mesh
- Step 2b: group terms by mesh-current
- Step 3: solve the system of equations

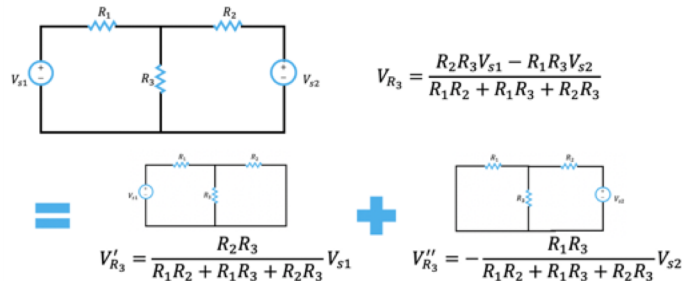
mesh current

a SUPERMESH is formed when a current source is connected in a path



A CIRCUIT is LINEAR  
if it possesses the properties  
of additive and homogeneity

Superposition  
trades off the examination  
of several simpler circuits in  
place of one complex circuit  
LINEAR CIRCUIT ONLY



Consider one INDEPENDENT source  
at a time, by "turning off" all other  
INDEPENDENT sources

DEPENDENT sources are  
left intact because they are  
controlled by circuit variables

Circuit Analysis Skill

Kirchhoff's Law +

Linear v.s. Non-linear +

Time v.s. Frequency +

Feedback +

Additional skills +

step by step process

Step 1: write the circuit equations according to KCL/KVL

Step 2a: find the complementary solution

Step 2b: find the particular integral solution

Step 3: find initial V/I to solve the unknown

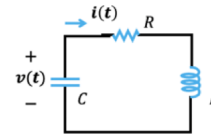
1st order examples

source free

with forcing func

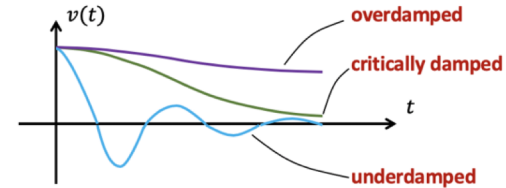
2nd order examples

source free



### A DISCUSSION ON $\zeta$

- If  $\zeta < 1 \rightarrow$  underdamped
- If  $\zeta = 1 \rightarrow$  critically damped
- If  $\zeta > 1 \rightarrow$  overdamped



Particular integral solution  
 $v_p(t) = 0$   
 Complementary solution  
 $v_c(t) = K_1 e^{s_1 t} + K_2 e^{s_2 t}$   
 Where  $\begin{cases} S_1 = -\zeta\omega_0 + \omega_0\sqrt{\zeta^2 - 1} \\ S_2 = -\zeta\omega_0 - \omega_0\sqrt{\zeta^2 - 1} \end{cases}$   
 and  $\omega_0 = \frac{1}{\sqrt{LC}}, \zeta = \frac{R}{2\omega_0 L}$   
 Full solution  $v(t) = v_p(t) + v_c(t)$

RLC circuit status

1. underdamped
2. critically damped
3. overdamped

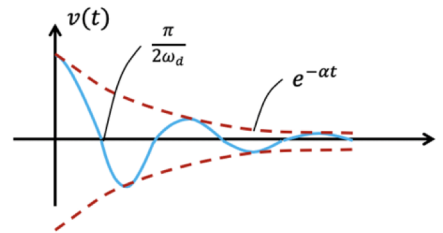
### Understand the waveform

$$v_c(t) = e^{-\zeta\omega_0 t} [A_1 \cos(\omega_d t) + A_2 \sin(\omega_d t)]$$

Define **DAMPING FACTOR**  $\alpha = \frac{R}{2L}$

Define **RESONATE FREQUENCY**  $\omega_0 = \frac{1}{\sqrt{LC}}$

Define **DAMPING RATIO**  $\zeta = \frac{\alpha}{\omega_0} = \frac{R}{2\omega_0 L}$



Transient v.s. Steady

Frequency Response

step by step process

- Step 1: write the circuit equations according to KCL/KVL
- Step 2a: find the complementary solution
- Step 2b: find the particular integral solution
- Step 3: find initial VII to solve the unknown

- 1st order examples
  - source free
  - with forcing func
- 2nd order examples
  - source free

**A DISCUSSION ON  $\zeta$**

- If  $\zeta < 1 \rightarrow$  underdamped
- If  $\zeta = 1 \rightarrow$  critically damped
- If  $\zeta > 1 \rightarrow$  overdamped

Particular integral solution  
 $v_p(t) = 0$

Complementary solution  
 $v_c(t) = K_1 e^{s_1 t} + K_2 e^{s_2 t}$

Where  $\begin{cases} s_1 = -\zeta\omega_0 + j\omega_0\sqrt{1-\zeta^2} \\ s_2 = -\zeta\omega_0 - j\omega_0\sqrt{1-\zeta^2} \end{cases}$

and  $\omega_0 = \frac{1}{\sqrt{LC}}$ ,  $\zeta = \frac{R}{2\omega_0 L}$

Full solution  $v(t) = v_p(t) + v_c(t)$

Understand the waveform  
 $v_c(t) = e^{-\zeta\omega_0 t} [A_1 \cos(\omega_d t) + A_2 \sin(\omega_d t)]$

Define **DAMPING FACTOR**  $\alpha = \frac{R}{2L}$

Define **RESONATE FREQUENCY**  $\omega_0 = \frac{1}{\sqrt{LC}}$

Define **DAMPING RATIO**  $\zeta = \frac{\alpha}{\omega_0} = \frac{R}{2\omega_0 L}$

- RLC circuit status
1. underdamped
  2. critically damped
  3. overdamped

TRANSFER FUNCTION of a circuit or system describes the output response to an input excitation as a function of the angular frequency

$$G(\omega) = \frac{V_{out}(\omega)}{V_{in}(\omega)}$$

Filters +

Quality factor -

Q is the ratio of energy stored to energy lost at the resonant frequency

$$Q = 2\pi \frac{W_S}{W_D}$$

Bode plot -

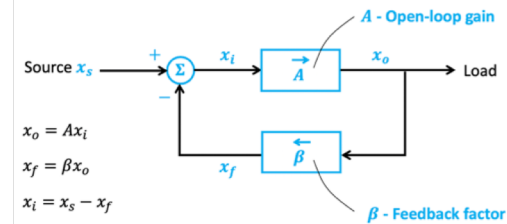
Bode plot generates a "straight-line" approximation of the transfer function

Basic rules +

Transient v.s. Steady +

Frequency Response +

### General negative feedback structure

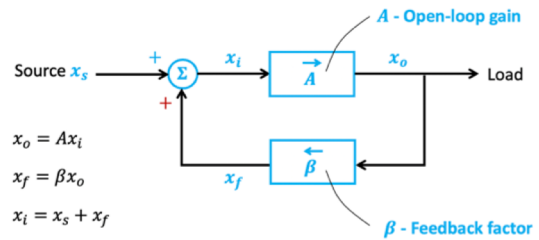


- The gain of the feedback amplifier  $A_f = \frac{x_o}{x_s} = \frac{A}{1 + A\beta}$ 
  - $1 + A\beta$  - amount of feedback
  - $A\beta$  - LOOP GAIN
- If  $A\beta \gg 1$   $A_f \approx \frac{1}{\beta}$  The gain of the feedback amplifier is almost determined by the feedback network!

Negative feedback

benefits from negative feedback

### General positive feedback structure



- The gain of the feedback amplifier  $A_f = \frac{x_o}{x_s} = \frac{A}{1 - A\beta}$   $A\beta$  is LOOP GAIN

Positive feedback

Kirchhoff's Law

Linear v.s. Non-linear

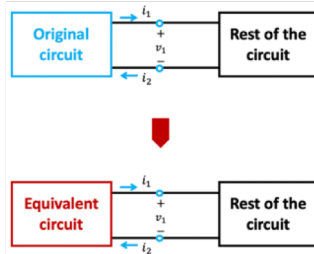
Time v.s. Frequency

Feedback

Additional skills

Circuit Analysis Skill

the general concept



resistors in series:  
sum of resistance

$$R_{eq} = \sum_{i=1}^N R_i$$

resistors in parallel:  
sum of conductance

$$\frac{1}{R_{eq}} = \sum_{i=1}^N \frac{1}{R_i}$$

w/o source

capacitors in series

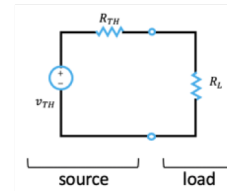
$$\frac{1}{C_{eq}} = \frac{1}{C_1} + \frac{1}{C_2} + \frac{1}{C_3}$$

capacitors in parallel

$$C_{eq} = C_1 + C_2 + C_3$$

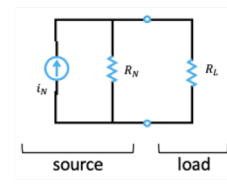
Thévenin's theorem

LINEAR two-terminal circuit can be replaced by an equivalent circuit composed of a voltage source and a series resistor



Norton's theorem

LINEAR two-terminal circuit can be replaced by an equivalent circuit composed of a current source and a parallel resistor



step by step process

to find Thévenin's/Norton's equivalent

Equivalent circuits

Source transformation

One-Port Network

Two-Port Network

ar ⊕

⊕

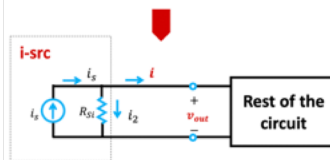
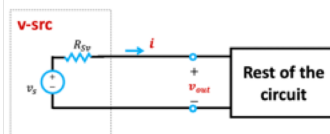
Equivalent circuits ⊕

Source transformation ⊕

One-Port Network ⊕

Two-Port Network ⊕

Goal: to turn a voltage source into current source, and vice-versa



- For the v-src, according to KVL

$$-v_s + iR_{Sv} + v_{out} = 0$$

- For the i-src, according to KCL

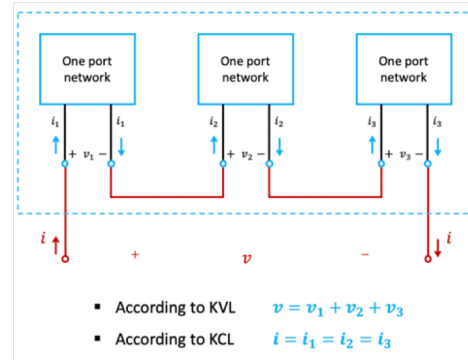
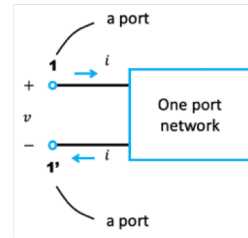
$$\begin{cases} i_s = i + i_2 \\ i_2 = \frac{v_{out}}{R_{Si}} \end{cases}$$

- If  $R_{Sv} = R_{Si} = R_S$

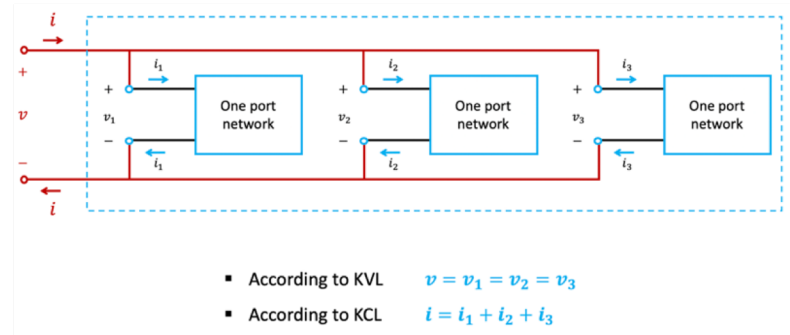
$$v_s = i_s R_S$$



ONE PORT NETWORK is a two terminal electrical network in which, current enters through one terminal and leaves through another terminal.



in series ⊖



in parallel ⊖

Connections ⊖

another way to find Thévenin / Norton's equivalent ⊖

- Step 1a: remove the load
- Step 1b: apply a test voltage to the port
- Step 2: find the relationship between  $v$ -test &  $i$ -test

- ar ⊕
- ⊕
- Equivalent circuits ⊕
- Source transformation ⊕
- One-Port Network ⊕
- Two-Port Network ⊕

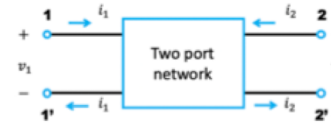
## TWO PORT NETWORK

is a pair of two terminal electrical network in which, current enters through one terminal and leaves through another terminal.

### PARAMETERS

If two of the four variables are independent and another two variables as dependent, the coefficients of the independent variables are called as PARAMETERS

Know how to calculate the network parameters



#### Z parameters

$$\begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} i_1 \\ i_2 \end{bmatrix}$$

#### T parameters

$$\begin{bmatrix} v_1 \\ i_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} v_2 \\ -i_2 \end{bmatrix}$$

#### h parameters

$$\begin{bmatrix} v_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} i_1 \\ v_2 \end{bmatrix}$$

#### Y parameters

$$\begin{bmatrix} i_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$$

#### T' parameters

$$\begin{bmatrix} v_2 \\ i_2 \end{bmatrix} = \begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} v_1 \\ -i_1 \end{bmatrix}$$

#### g parameters

$$\begin{bmatrix} i_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} g_{11} & g_{12} \\ g_{21} & g_{22} \end{bmatrix} \begin{bmatrix} v_1 \\ i_2 \end{bmatrix}$$

$$Z = Y^{-1}$$

$$T = T'^{-1}$$

$$h = g^{-1}$$

- Connections
- in series ⊕
  - in parallel ⊕
  - cascading ⊕

Equivalent circuits ⊕

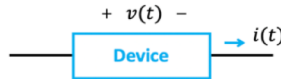
Source transformation ⊕

One-Port Network ⊕

Two-Port Network ⊕

# Foundations of Electronic Circuits & Systems

## Physical Quantity



- Voltage  $V$  ⊕
- Current  $I$  ⊕
- Power  $P$  ⊕
- Energy  $w$  ⊕
- Complex power ⊕
- Impedance ⊕
- s-domain model ⊕

## Electronic Devices

- Linear v.s. Non-linear ⊕
- passive v.s. active ⊕

## Circuit Analysis Skill

- Kirchhoff's Law ⊕
- Linear v.s. Non-linear ⊕
- Time v.s. Frequency ⊕
- Feedback ⊕
- Additional skills ⊕

# Good Luck

